

# NCP163

## 250 mA, Ultra-Low Noise and High PSRR LDO Regulator for RF and Analog Circuits

The NCP163 is a next generation of high PSRR, ultra-low noise LDO capable of supplying 250 mA output current. Designed to meet the requirements of RF and sensitive analog circuits, the NCP163 device provides ultra-low noise, high PSRR and low quiescent current. The device also offer excelent load/line transients. The NCP163 is designed to work with a 1  $\mu$ F input and a 1  $\mu$ F output ceramic capacitor. It is available in two thickness ultra-small 0.35P, 0.65 mm x 0.65 mm Chip Scale Package (CSP) and XDFN4 0.65P, 1 mm x 1 mm.

### Features

- Operating Input Voltage Range: 2.2 V to 5.5 V
- Available in Fixed Voltage Option: 1.2 V to 5.3 V
- $\pm 2\%$  Accuracy Over Load/Temperature
- Ultra Low Quiescent Current Typ. 12  $\mu$ A
- Standby Current: Typ. 0.1  $\mu$ A
- Very Low Dropout: 80 mV at 250 mA
- Ultra High PSRR: Typ. 92 dB at 20 mA,  $f = 1$  kHz
- Ultra Low Noise: 6.5  $\mu$ V<sub>RMS</sub>
- Stable with a 1  $\mu$ F Small Case Size Ceramic Capacitors
- Available in –WLCSP4 0.65 mm x 0.65 mm x 0.33 mm  
–WLCSP4 0.65 mm x 0.65 mm x 0.4 mm  
–XDFN4 1 mm x 1 mm x 0.4 mm
- These Devices are Pb-Free, Halogen Free/BFR Free and are RoHS Compliant

### Typical Applications

- Battery-powered Equipment
- Wireless LAN Devices
- Smartphones, Tablets
- Cameras, DVRs, STB and Camcorders

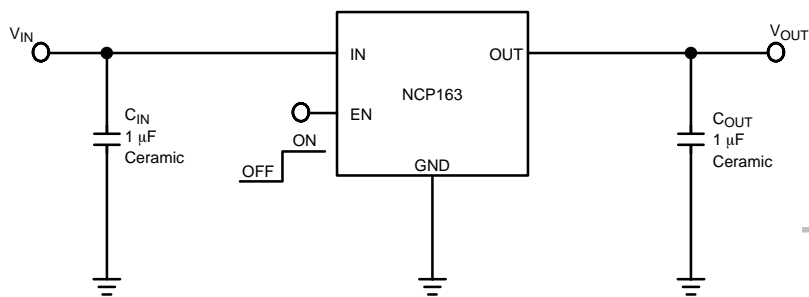


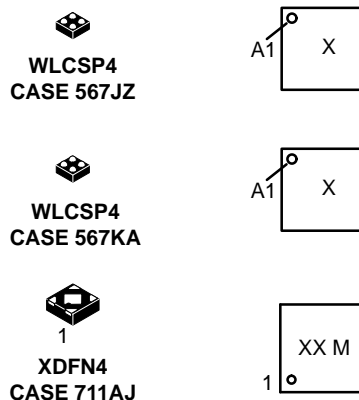
Figure 1. Typical Application Schematics



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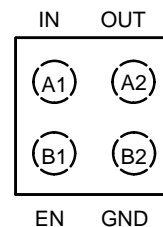
[www.onsemi.com](http://www.onsemi.com)

### MARKING DIAGRAMS

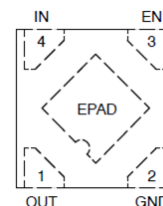


X or XX = Specific Device Code  
M = Date Code

### PIN CONNECTIONS



(Top View)



(Top View)

### ORDERING INFORMATION

See detailed ordering, marking and shipping information on page 15 of this data sheet.

# NCP163

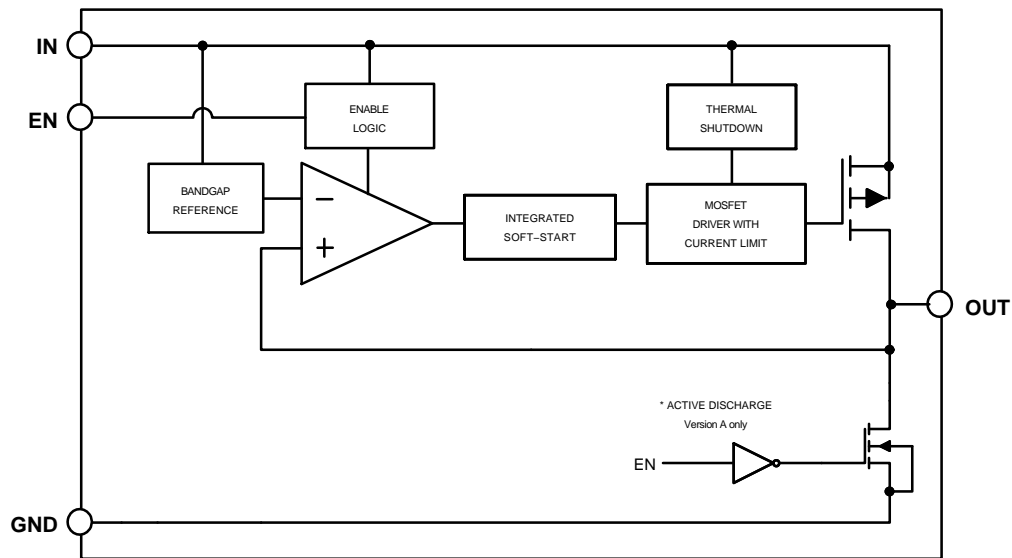


Figure 2. Simplified Schematic Block Diagram

## PIN FUNCTION DESCRIPTION

Pin No. WLCSP4	Pin No. XDFN4	Pin Name	Description
A1	4	IN	Input voltage supply pin
A2	1	OUT	Regulated output voltage. The output should be bypassed with small 1 $\mu$ F ceramic capacitor.
B1	3	EN	Chip enable: Applying $V_{EN} < 0.4$ V disables the regulator, Pulling $V_{EN} > 1.2$ V enables the LDO.
B2	2	GND	Common ground connection
-	EPAD	EPAD	Expose pad can be tied to ground plane for better power dissipation

## ABSOLUTE MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Input Voltage (Note 1)	$V_{IN}$	-0.3 V to 6	V
Output Voltage	$V_{OUT}$	-0.3 to $V_{IN} + 0.3$ , max. 6 V	V
Chip Enable Input	$V_{CE}$	-0.3 to $V_{IN} + 0.3$ , max. 6 V	V
Output Short Circuit Duration	$t_{SC}$	unlimited	s
Maximum Junction Temperature	$T_J$	150	$^{\circ}$ C
Storage Temperature	$T_{STG}$	-55 to 150	$^{\circ}$ C
ESD Capability, Human Body Model (Note 2)	$ESD_{HBM}$	2000	V
ESD Capability, Machine Model (Note 2)	$ESD_{MM}$	200	V
ESD Capability, Charged Device Model (Note 2)	$ESD_{CDM}$	1000	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.

2. This device series incorporates ESD protection and is tested by the following methods:

ESD Human Body Model tested per EIA/JESD22-A114

ESD Machine Model tested per EIA/JESD22-A115

ESD Charged Device Model tested per EIA/JESD22-C101, Field Induced Charge Model

Latchup Current Maximum Rating tested per JEDEC standard: JESD78.

## THERMAL CHARACTERISTICS

Rating	Symbol	Value	Unit
Thermal Characteristics, WLCSP4 (Note 3), Thermal Resistance, Junction-to-Air	$R_{\theta JA}$	108	$^{\circ}$ C/W
Thermal Characteristics, XDFN4 (Note 3), Thermal Resistance, Junction-to-Air		198.1	

3. Measured according to JEDEC board specification. Detailed description of the board can be found in JESD51-7

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**ELECTRICAL CHARACTERISTICS**  $-40^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$ ;  $V_{IN} = V_{OUT(NOM)} + 1\text{ V}$ ;  $I_{OUT} = 1\text{ mA}$ ,  $C_{IN} = C_{OUT} = 1\text{ }\mu\text{F}$ , unless otherwise noted.  $V_{EN} = 1.2\text{ V}$ . Typical values are at  $T_J = +25^{\circ}\text{C}$  (Note 4).

Parameter	Test Conditions	Symbol	Min	Typ	Max	Unit
Operating Input Voltage		$V_{IN}$	2.2		5.5	V
Output Voltage Accuracy	$V_{IN} = (V_{OUT(NOM)} + 1\text{ V})$ to 5.5 V $0\text{ mA} \leq I_{OUT} \leq 250\text{ mA}$	$V_{OUT}$	-2		+2	%
	$V_{IN} = (V_{OUT(NOM)} + 1\text{ V})$ to 5.5 V $0\text{ mA} \leq I_{OUT} \leq 250\text{ mA}$ (for $V_{OUT} < 1.8\text{ V}$ , XDFN4 package)	$V_{OUT}$	-3		+3	%
Line Regulation	$V_{OUT(NOM)} + 1\text{ V} \leq V_{IN} \leq 5.5\text{ V}$	$Line_{Reg}$		0.02		%/V
Load Regulation	$I_{OUT} = 1\text{ mA}$ to 250 mA	$Load_{Reg}$		0.001		%/mA
Dropout Voltage (Note 5)	$I_{OUT} = 250\text{ mA}$	$V_{DO}$		180	250	mV
				110	175	
				95	160	
				90	155	
				85	149	
				80	145	
				75	140	
				65	120	
			75	105		
Output Current Limit	$V_{OUT} = 90\% V_{OUT(NOM)}$	$I_{CL}$	250	700		mA
Short Circuit Current	$V_{OUT} = 0\text{ V}$	$I_{SC}$		690		
Quiescent Current	$I_{OUT} = 0\text{ mA}$	$I_Q$		12	20	$\mu\text{A}$
Shutdown Current	$V_{EN} \leq 0.4\text{ V}$ , $V_{IN} = 4.8\text{ V}$	$I_{DIS}$		0.01	1	$\mu\text{A}$
EN Pin Threshold Voltage	EN Input Voltage "H"	$V_{ENH}$	1.2			V
	EN Input Voltage "L"	$V_{ENL}$			0.4	
EN Pull Down Current	$V_{EN} = 4.8\text{ V}$	$I_{EN}$		0.2	0.5	$\mu\text{A}$
Turn-On Time	$C_{OUT} = 1\text{ }\mu\text{F}$ , From assertion of $V_{EN}$ to $V_{OUT} = 95\% V_{OUT(NOM)}$			120		$\mu\text{s}$
Power Supply Rejection Ratio	$I_{OUT} = 20\text{ mA}$	$PSRR$	$f = 100\text{ Hz}$	91		dB
	$f = 1\text{ kHz}$		92			
	$f = 10\text{ kHz}$		85			
	$f = 100\text{ kHz}$		60			
Output Voltage Noise	$f = 10\text{ Hz}$ to 100 kHz	$V_N$		8.0		$\mu\text{V}_{RMS}$
				6.5		
Thermal Shutdown Threshold	Temperature rising	$T_{SDH}$		160		$^{\circ}\text{C}$
	Temperature falling	$T_{SDL}$		140		$^{\circ}\text{C}$
Active Output Discharge Resistance	$V_{EN} < 0.4\text{ V}$ , Version A only	$R_{DIS}$		280		$\Omega$
Line Transient (Note 6)	$V_{IN} = (V_{OUT(NOM)} + 1\text{ V})$ to $(V_{OUT(NOM)} + 1.6\text{ V})$ in 30 $\mu\text{s}$ , $I_{OUT} = 1\text{ mA}$	$Tran_{LINE}$	-1			mV
	$V_{IN} = (V_{OUT(NOM)} + 1.6\text{ V})$ to $(V_{OUT(NOM)} + 1\text{ V})$ in 30 $\mu\text{s}$ , $I_{OUT} = 1\text{ mA}$				+1	
Load Transient (Note 6)	$I_{OUT} = 1\text{ mA}$ to 200 mA in 10 $\mu\text{s}$	$Tran_{LOAD}$	-40			mV
	$I_{OUT} = 200\text{ mA}$ to 1mA in 10 $\mu\text{s}$				+40	

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

4. Performance guaranteed over the indicated operating temperature range by design and/or characterization. Production tested at  $T_A = 25^{\circ}\text{C}$ . Low duty cycle pulse techniques are used during the testing to maintain the junction temperature as close to ambient as possible.
5. Dropout voltage is characterized when  $V_{OUT}$  falls 100 mV below  $V_{OUT(NOM)}$ .
6. Guaranteed by design.

TYPICAL CHARACTERISTICS

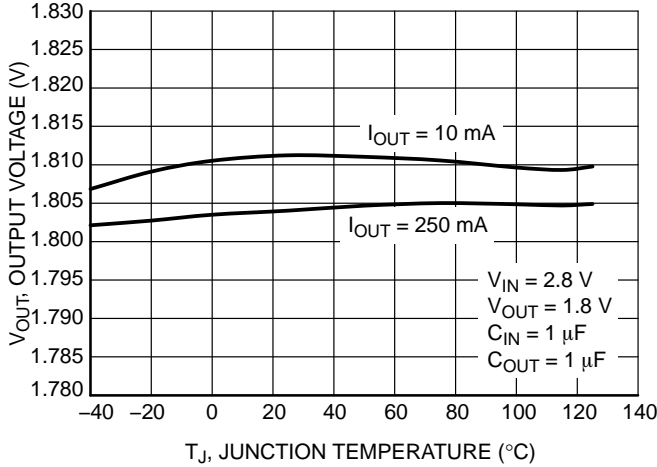


Figure 3. Output Voltage vs. Temperature –  $V_{OUT} = 1.8\text{ V}$  – XDFN Package

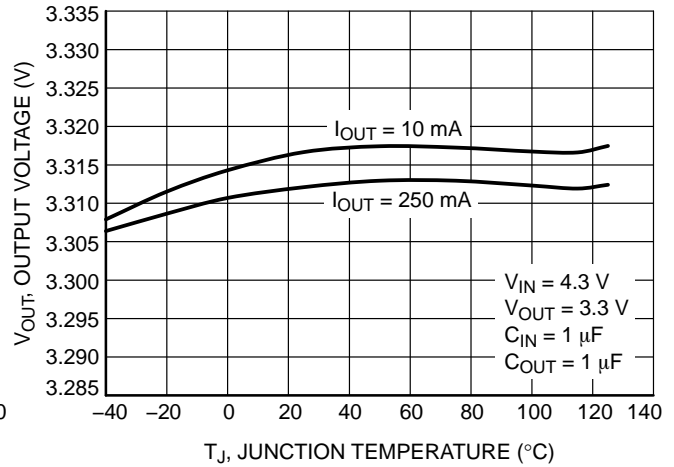


Figure 4. Output Voltage vs. Temperature –  $V_{OUT} = 3.3\text{ V}$  – XDFN Package

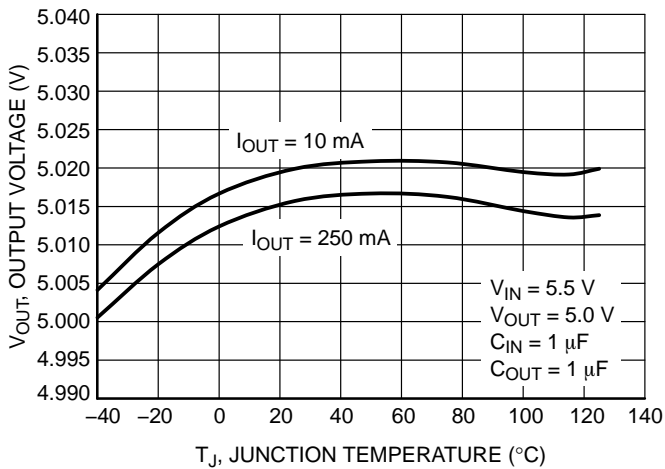


Figure 5. Output Voltage vs. Temperature –  $V_{OUT} = 5.0\text{ V}$  – XDFN Package

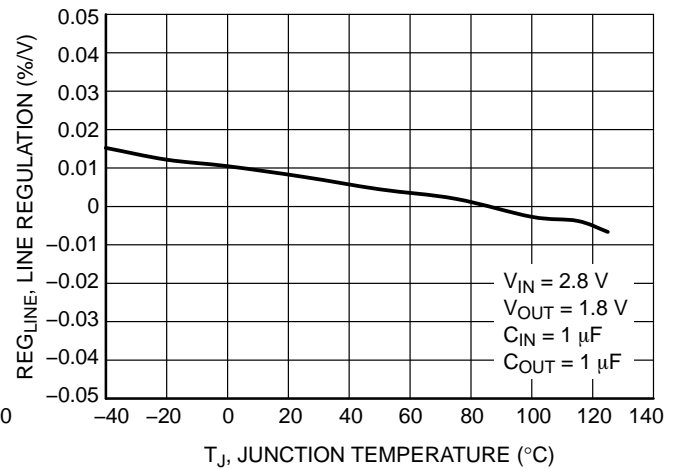


Figure 6. Line Regulation vs. Temperature –  $V_{OUT} = 1.8\text{ V}$

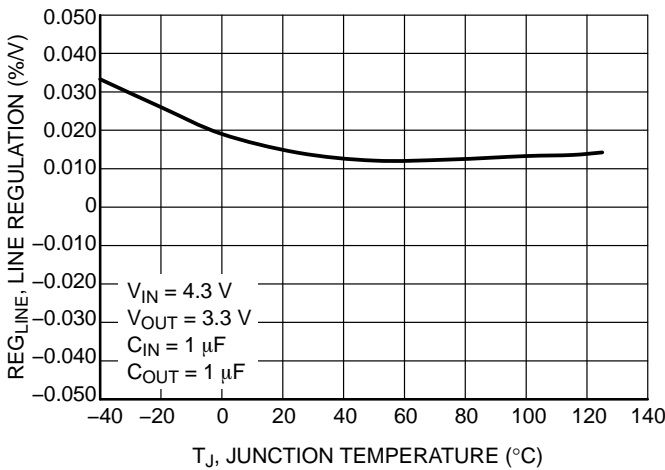


Figure 7. Line Regulation vs. Temperature –  $V_{OUT} = 3.3\text{ V}$

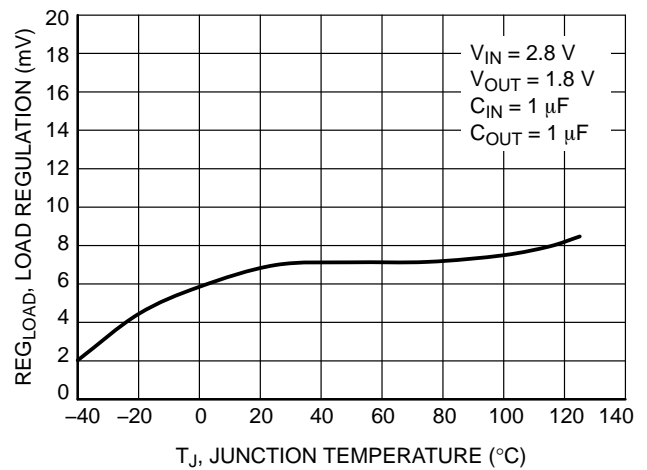


Figure 8. Load Regulation vs. Temperature –  $V_{OUT} = 1.8\text{ V}$

TYPICAL CHARACTERISTICS

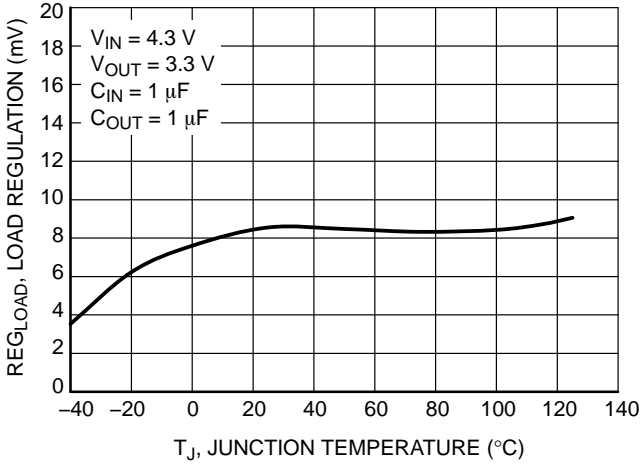


Figure 9. Load Regulation vs. Temperature –  $V_{OUT} = 3.3\text{ V}$

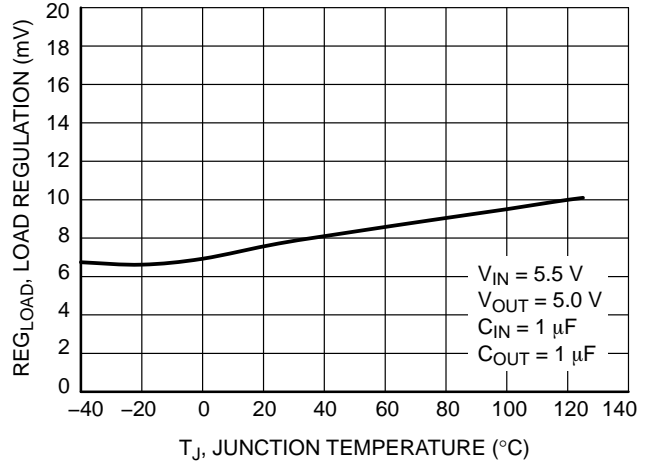


Figure 10. Load Regulation vs. Temperature –  $V_{OUT} = 5.0\text{ V}$

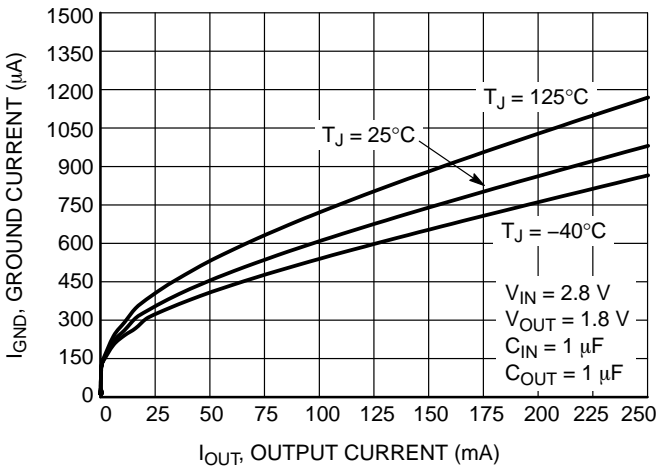


Figure 11. Ground Current vs. Load Current –  $V_{OUT} = 1.8\text{ V}$

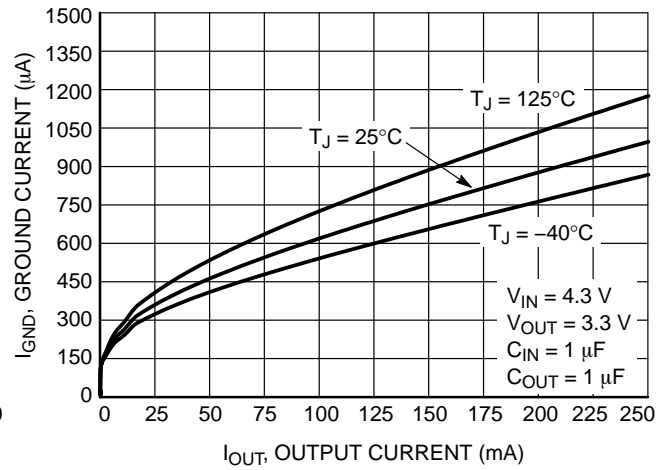


Figure 12. Ground Current vs. Load Current –  $V_{OUT} = 3.3\text{ V}$

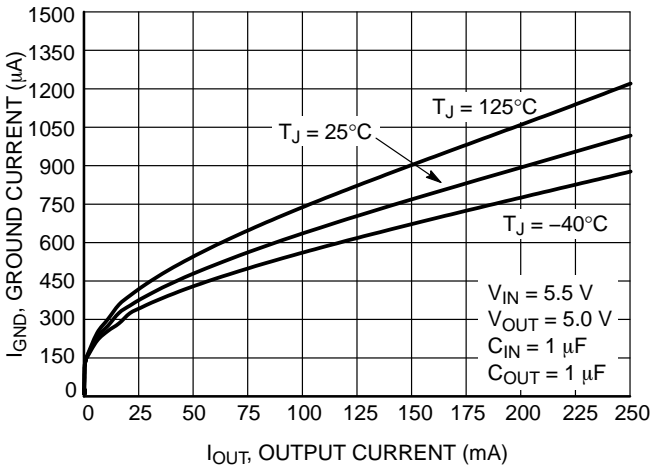


Figure 13. Ground Current vs. Load Current –  $V_{OUT} = 5.0\text{ V}$

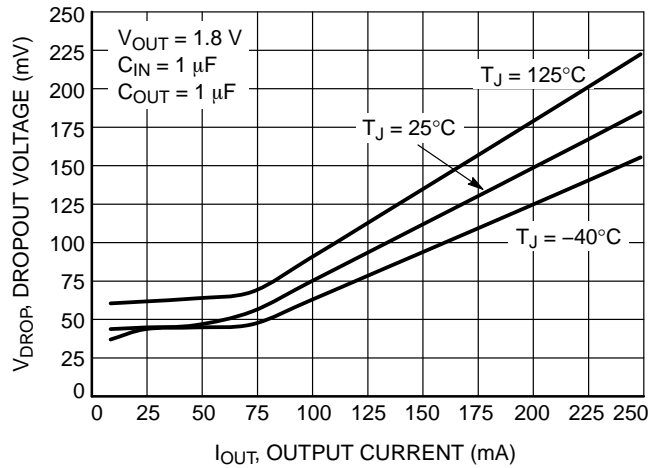


Figure 14. Dropout Voltage vs. Load Current –  $V_{OUT} = 1.8\text{ V}$

TYPICAL CHARACTERISTICS

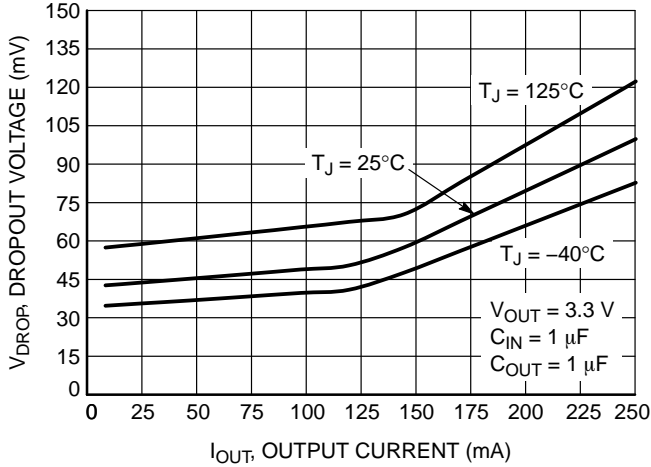


Figure 15. Dropout Voltage vs. Load Current –  $V_{OUT} = 3.3\text{ V}$

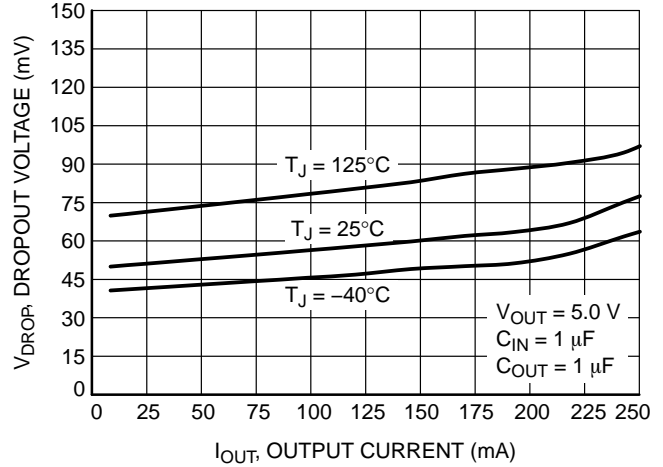


Figure 16. Dropout Voltage vs. Load Current –  $V_{OUT} = 5.0\text{ V}$

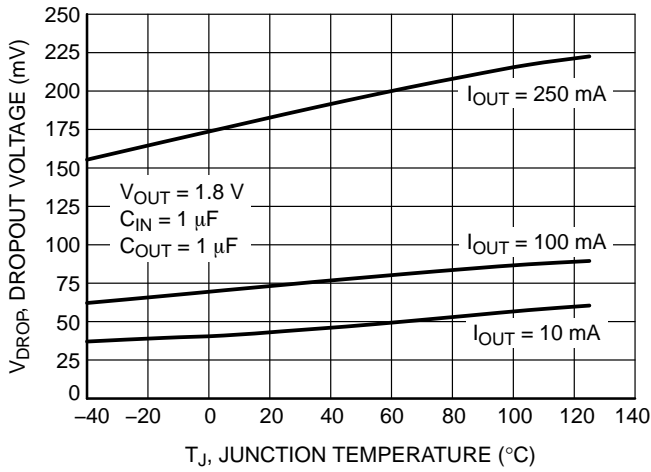


Figure 17. Dropout Voltage vs. Temperature –  $V_{OUT} = 1.8\text{ V}$

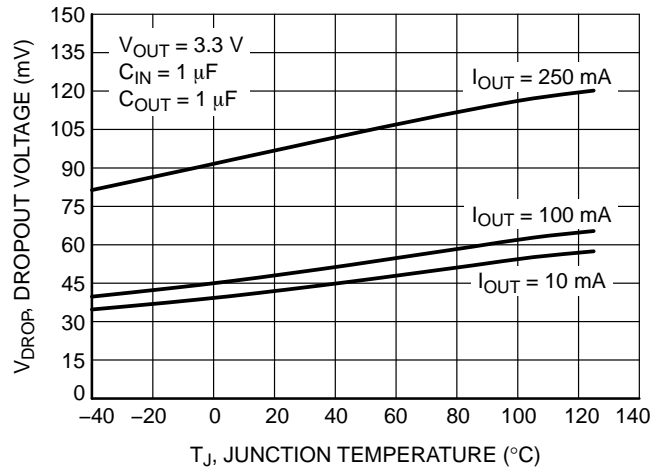


Figure 18. Dropout Voltage vs. Temperature –  $V_{OUT} = 3.3\text{ V}$

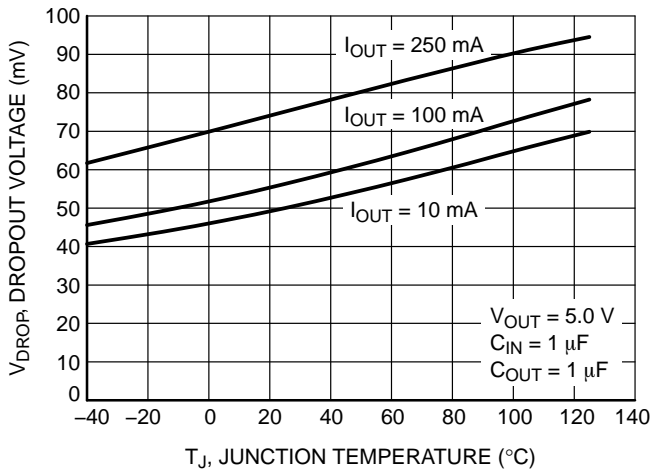


Figure 19. Dropout Voltage vs. Temperature –  $V_{OUT} = 5.0\text{ V}$

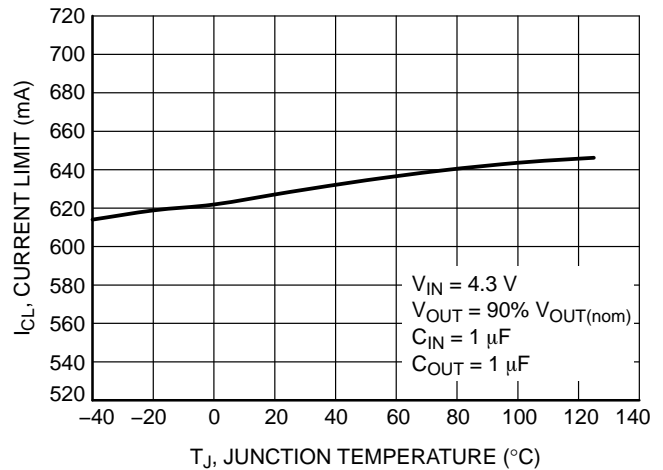


Figure 20. Current Limit vs. Temperature

TYPICAL CHARACTERISTICS

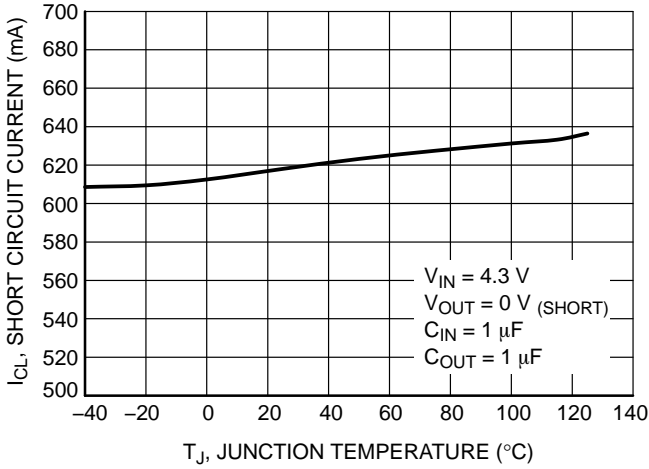


Figure 21. Short Circuit Current vs. Temperature

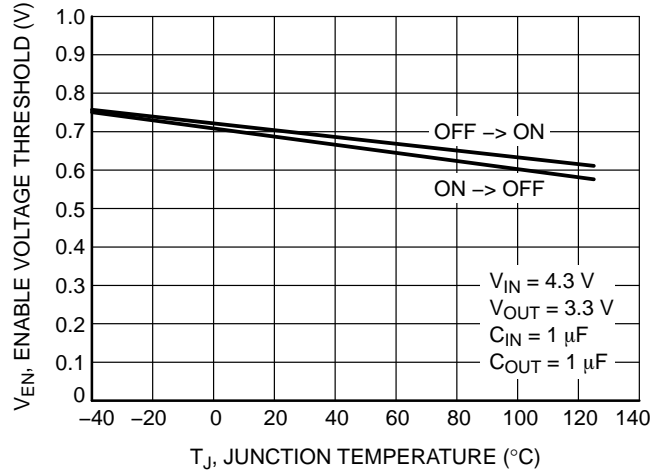


Figure 22. Enable Thresholds Voltage

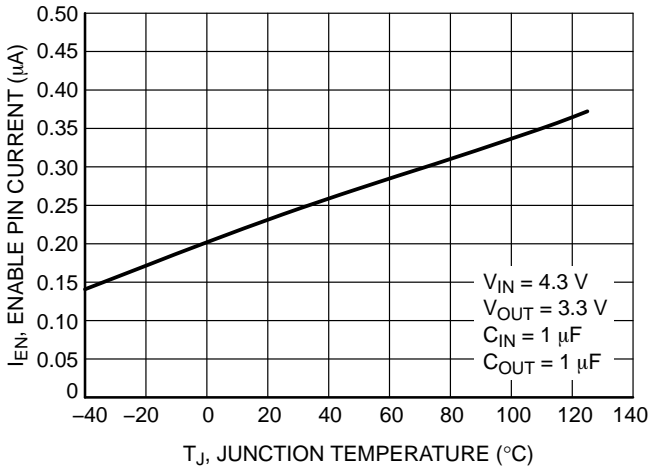


Figure 23. Current to Enable Pin vs. Temperature

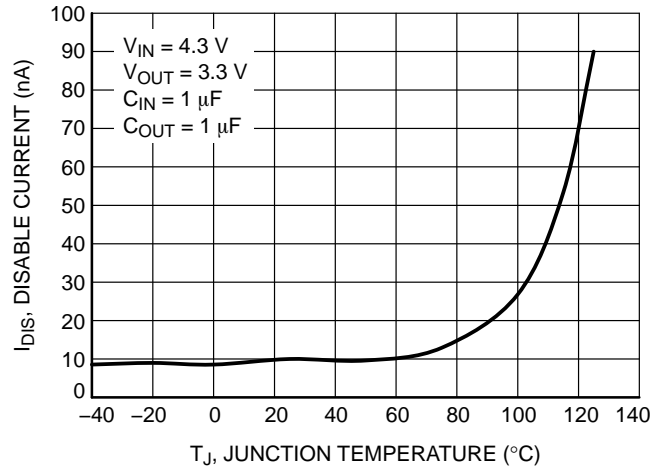


Figure 24. Disable Current vs. Temperature

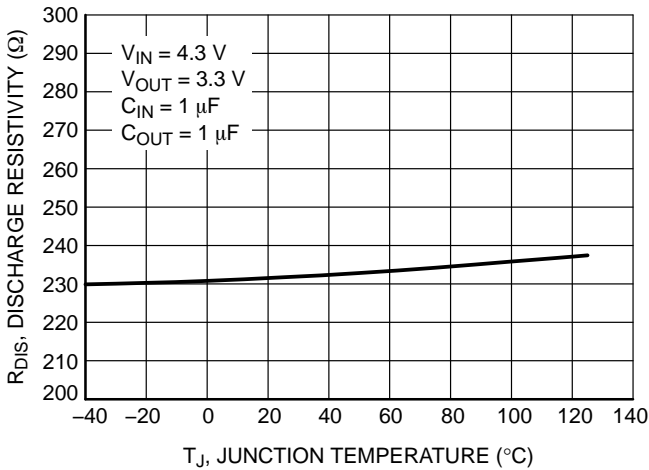


Figure 25. Discharge Resistance vs. Temperature

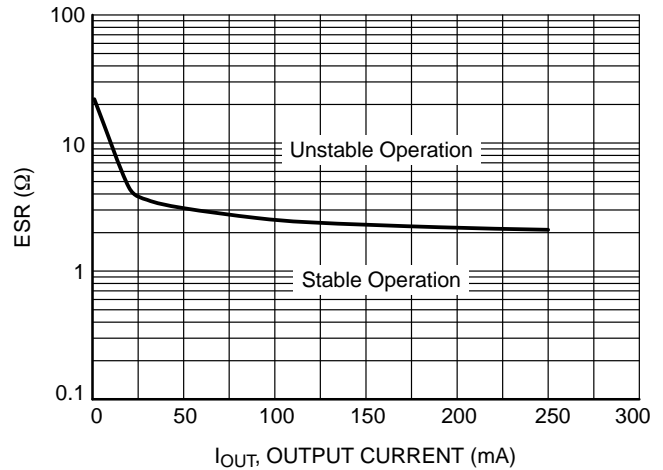
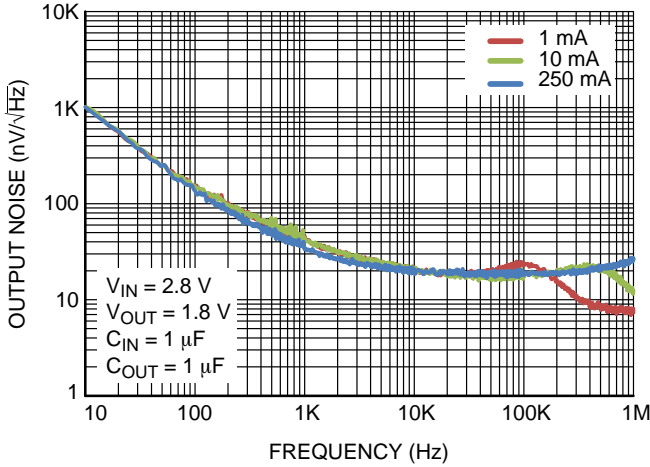


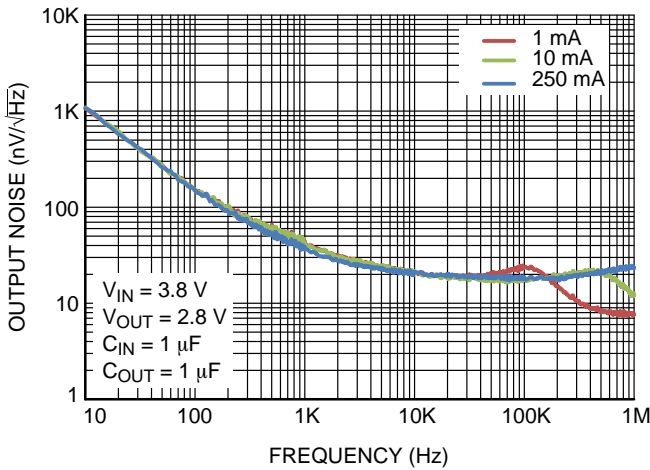
Figure 26. Maximum C<sub>OUT</sub> ESR Value vs. Load Current

TYPICAL CHARACTERISTICS



$I_{OUT}$	RMS Output Noise ( $\mu\text{V}$ )	
	10 Hz – 100 kHz	100 Hz – 100 kHz
1 mA	7.73	6.99
10 mA	7.12	6.26
250 mA	7.11	6.33

Figure 27. Output Voltage Noise Spectral Density –  $V_{OUT} = 1.8\text{ V}$



$I_{OUT}$	RMS Output Noise ( $\mu\text{V}$ )	
	10 Hz – 100 kHz	100 Hz – 100 kHz
1 mA	7.9	7.07
10 mA	7.19	6.25
250 mA	7.29	6.38

Figure 28. Output Voltage Noise Spectral Density –  $V_{OUT} = 2.8\text{ V}$

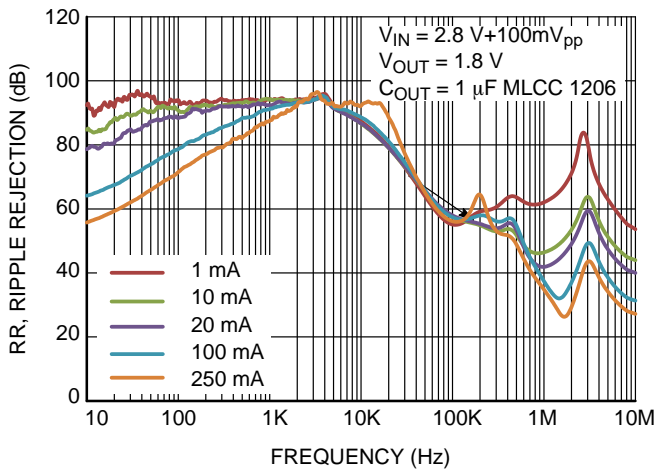


Figure 29. Power Supply Rejection Ratio –  $V_{OUT} = 1.8\text{ V}$

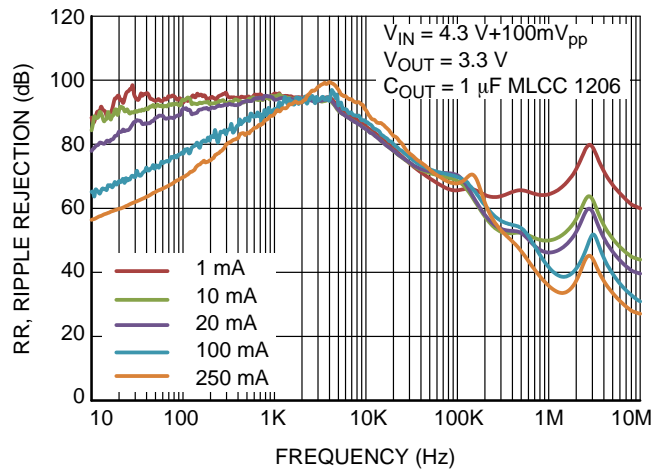


Figure 30. Power Supply Rejection Ratio –  $V_{OUT} = 3.3\text{ V}$



TYPICAL CHARACTERISTICS

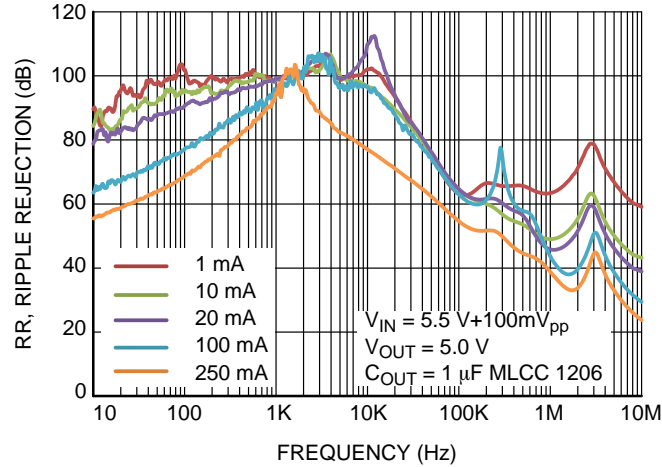


Figure 31. Power Supply Rejection Ratio –  
 $V_{OUT} = 5.0\text{ V}$

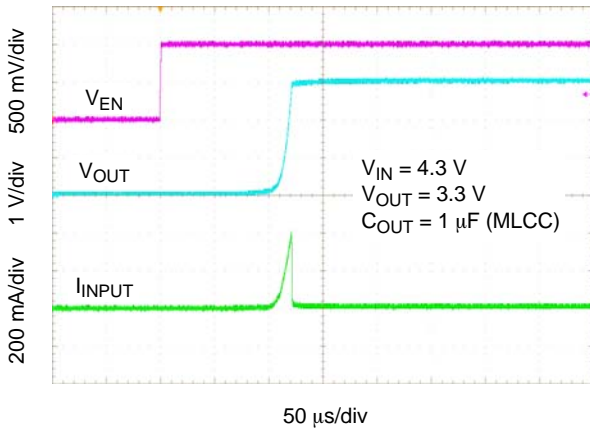


Figure 32. Enable Turn-on Response –  
 $C_{OUT} = 1\ \mu\text{F}$ ,  $I_{OUT} = 10\ \text{mA}$

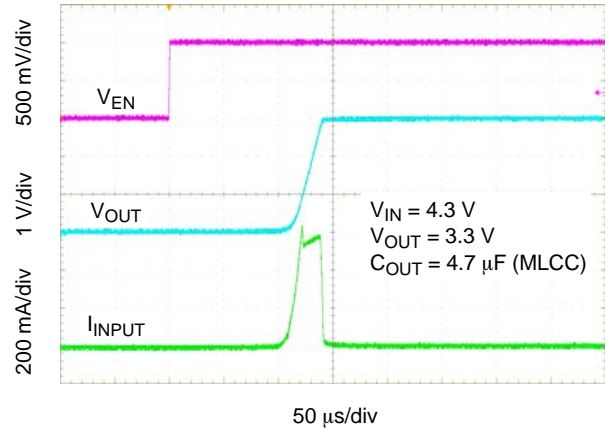


Figure 33. Enable Turn-on Response –  
 $C_{OUT} = 4.7\ \mu\text{F}$ ,  $I_{OUT} = 10\ \text{mA}$

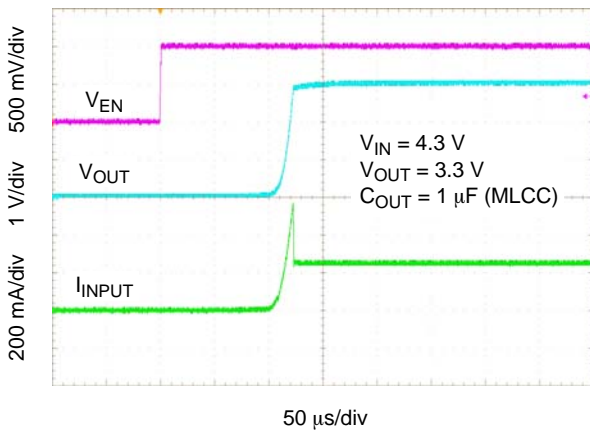


Figure 34. Enable Turn-on Response –  
 $C_{OUT} = 1\ \mu\text{F}$ ,  $I_{OUT} = 250\ \text{mA}$

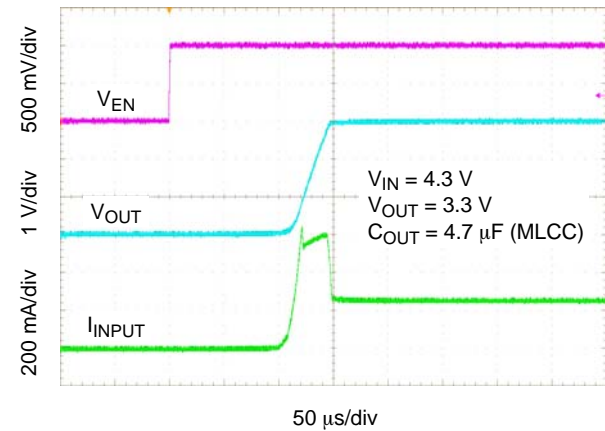


Figure 35. Enable Turn-on Response –  
 $C_{OUT} = 4.7\ \mu\text{F}$ ,  $I_{OUT} = 250\ \text{mA}$

TYPICAL CHARACTERISTICS

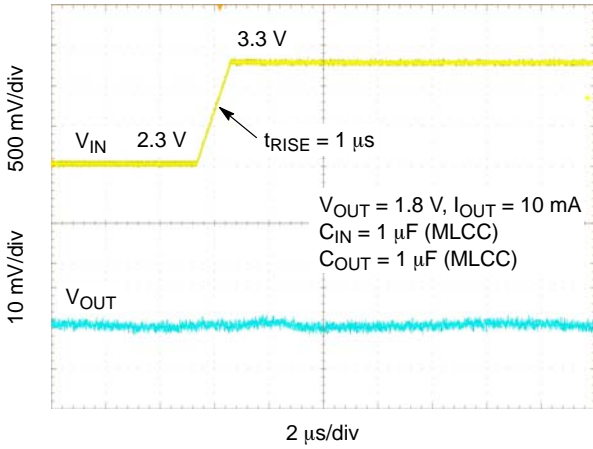


Figure 36. Line Transient Response –  
I<sub>OUT</sub> = 10 mA

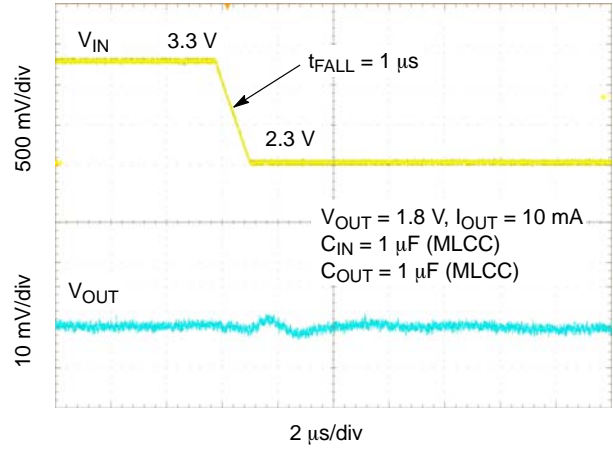


Figure 37. Line Transient Response –  
I<sub>OUT</sub> = 10 mA

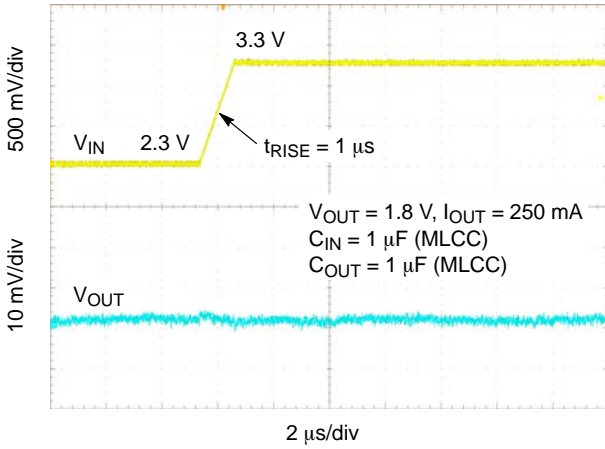


Figure 38. Line Transient Response –  
I<sub>OUT</sub> = 250 mA

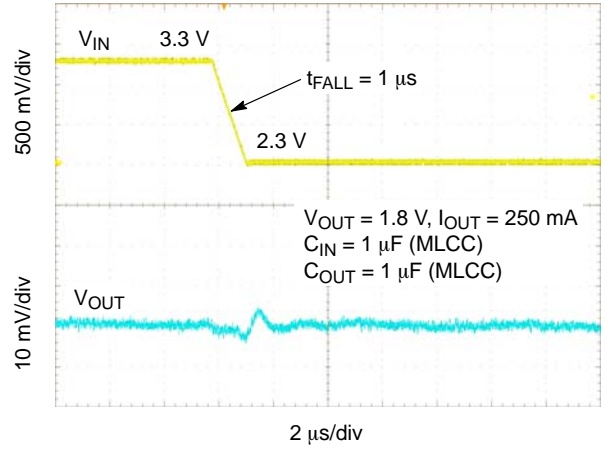


Figure 39. Line Transient Response –  
I<sub>OUT</sub> = 250 mA

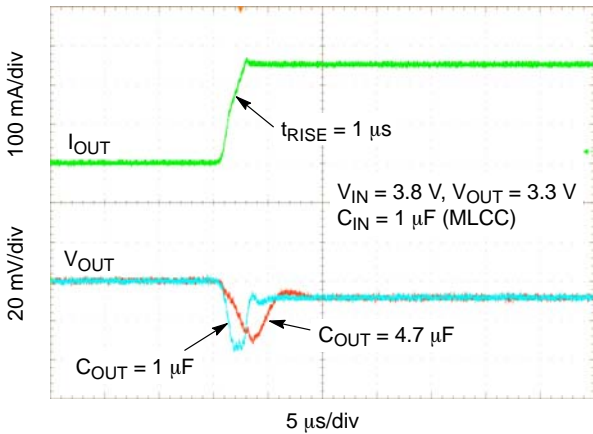


Figure 40. Load Transient Response –  
1 mA to 250 mA

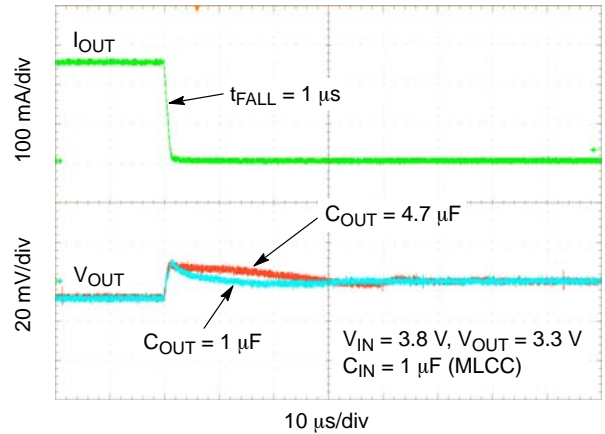


Figure 41. Load Transient Response –  
250 mA to 1 mA

TYPICAL CHARACTERISTICS

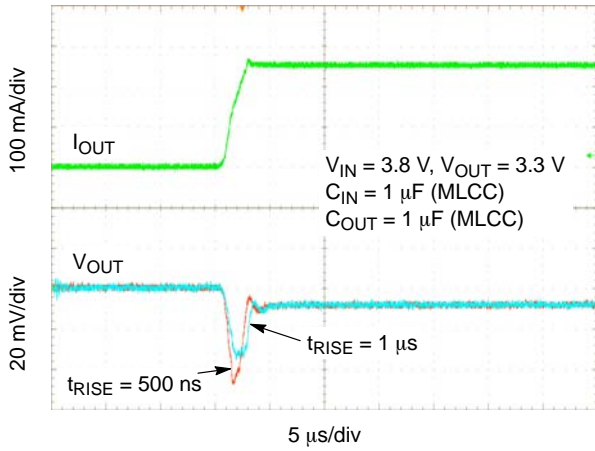


Figure 42. Load Transient Response – 1 mA to 250 mA

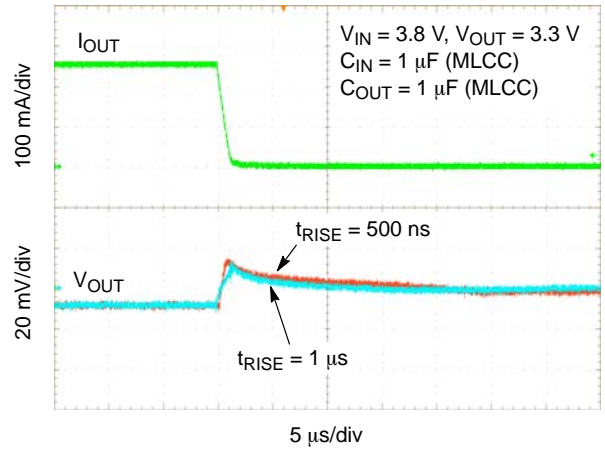


Figure 43. Load Transient Response – 250 mA to 1 mA

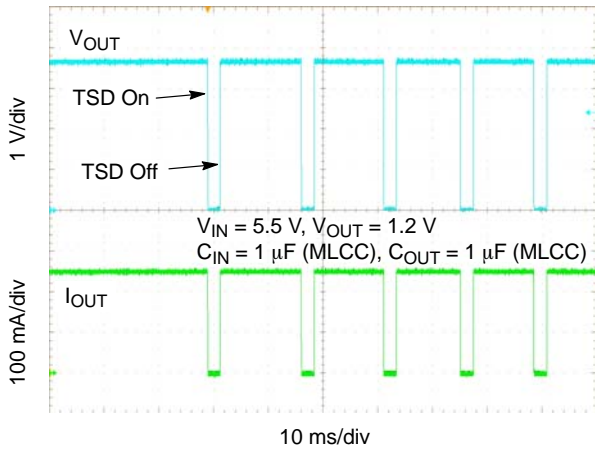


Figure 44. Overheating Protection – TSD

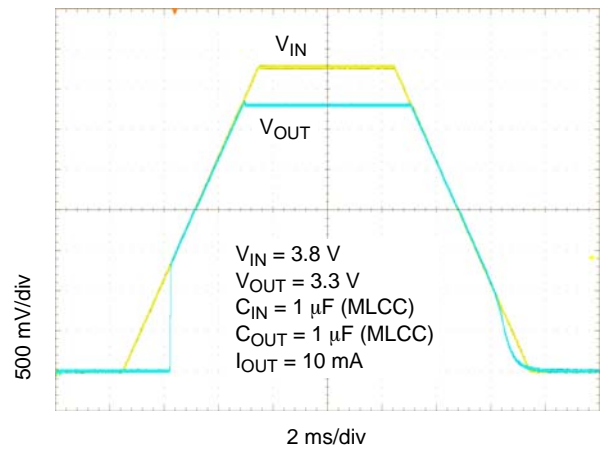


Figure 45. Turn-on/off – Slow Rising VIN

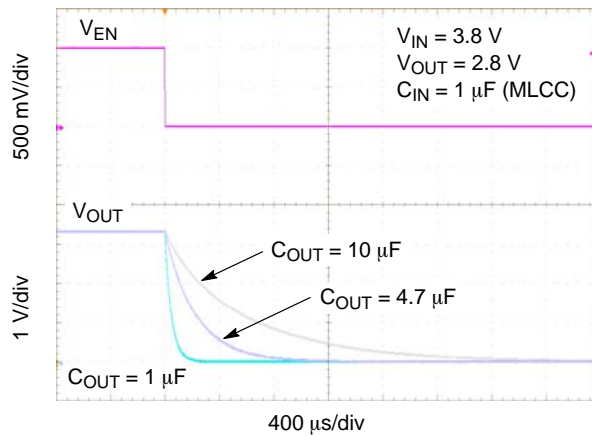


Figure 46. Enable Turn-off – Various Output Capacitors

APPLICATIONS INFORMATION

**General**

The NCP163 is an ultra–low noise 250 mA low dropout regulator designed to meet the requirements of RF applications and high performance analog circuits. The NCP163 device provides very high PSRR and excellent dynamic response. In connection with low quiescent current this device is well suitable for battery powered application such as cell phones, tablets and other. The NCP163 is fully protected in case of current overload, output short circuit and overheating.

**Input Capacitor Selection (C<sub>IN</sub>)**

Input capacitor connected as close as possible is necessary for ensure device stability. The X7R or X5R capacitor should be used for reliable performance over temperature range. The value of the input capacitor should be 1 μF or greater to ensure the best dynamic performance. This capacitor will provide a low impedance path for unwanted AC signals or noise modulated onto constant input voltage. There is no requirement for the ESR of the input capacitor but it is recommended to use ceramic capacitors for their low ESR and ESL. A good input capacitor will limit the influence of input trace inductance and source resistance during sudden load current changes.

**Output Decoupling (C<sub>OUT</sub>)**

The NCP163 requires an output capacitor connected as close as possible to the output pin of the regulator. The recommended capacitor value is 1 μF and X7R or X5R dielectric due to its low capacitance variations over the specified temperature range. The NCP163 is designed to remain stable with minimum effective capacitance of 0.7 μF to account for changes with temperature, DC bias and package size. Especially for small package size capacitors such as 0201 the effective capacitance drops rapidly with the applied DC bias. Please refer Figure 47.

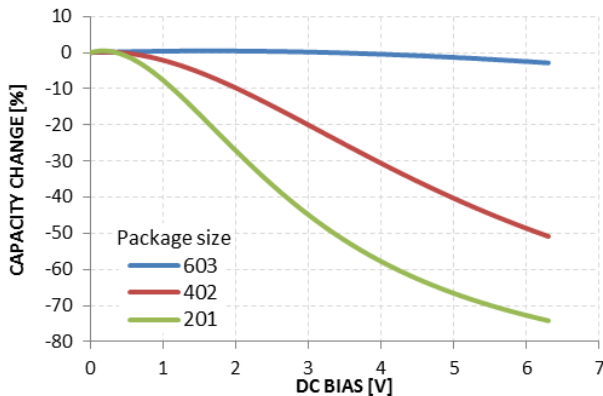


Figure 47. Capacity vs DC Bias Voltage

There is no requirement for the minimum value of Equivalent Series Resistance (ESR) for the C<sub>OUT</sub> but the maximum value of ESR should be less than 2 Ω. Larger output capacitors and lower ESR could improve the load

transient response or high frequency PSRR. It is not recommended to use tantalum capacitors on the output due to their large ESR. The equivalent series resistance of tantalum capacitors is also strongly dependent on the temperature, increasing at low temperature.

**Enable Operation**

The NCP163 uses the EN pin to enable/disable its device and to deactivate/activate the active discharge function.

If the EN pin voltage is <0.4 V the device is guaranteed to be disabled. The pass transistor is turned–off so that there is virtually no current flow between the IN and OUT. The active discharge transistor is active so that the output voltage V<sub>OUT</sub> is pulled to GND through a 280 Ω resistor. In the disable state the device consumes as low as typ. 10 nA from the V<sub>IN</sub>.

If the EN pin voltage >1.2 V the device is guaranteed to be enabled. The NCP163 regulates the output voltage and the active discharge transistor is turned–off.

The EN pin has internal pull–down current source with typ. value of 200 nA which assures that the device is turned–off when the EN pin is not connected. In the case where the EN function isn’t required the EN should be tied directly to IN.

**Output Current Limit**

Output Current is internally limited within the IC to a typical 700 mA. The NCP163 will source this amount of current measured with a voltage drops on the 90% of the nominal V<sub>OUT</sub>. If the Output Voltage is directly shorted to ground (V<sub>OUT</sub> = 0 V), the short circuit protection will limit the output current to 690 mA (typ). The current limit and short circuit protection will work properly over whole temperature range and also input voltage range. There is no limitation for the short circuit duration.

**Thermal Shutdown**

When the die temperature exceeds the Thermal Shutdown threshold (T<sub>SD</sub> – 160°C typical), Thermal Shutdown event is detected and the device is disabled. The IC will remain in this state until the die temperature decreases below the Thermal Shutdown Reset threshold (T<sub>SDR</sub> – 140°C typical). Once the IC temperature falls below the 140°C the LDO is enabled again. The thermal shutdown feature provides the protection from a catastrophic device failure due to accidental overheating. This protection is not intended to be used as a substitute for proper heat sinking.

**Power Dissipation**

As power dissipated in the NCP163 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the

# NCP163

ambient temperature affect the rate of junction temperature rise for the part.

The maximum power dissipation the NCP163 can handle is given by:

$$P_{D(MAX)} = \frac{[125^{\circ}\text{C} - T_A]}{\theta_{JA}} \quad (\text{eq. 1})$$

The power dissipated by the NCP163 for given application conditions can be calculated from the following equations:

$$P_D \approx V_{IN} \cdot I_{GND} + I_{OUT}(V_{IN} - V_{OUT}) \quad (\text{eq. 2})$$

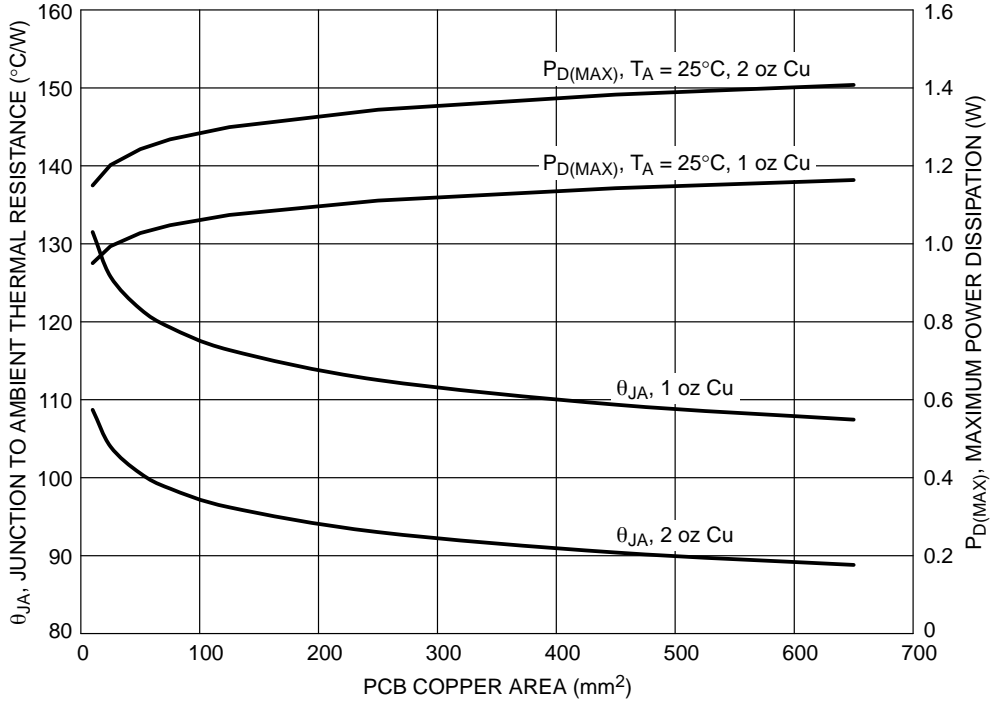


Figure 48.  $\theta_{JA}$  and  $P_{D(MAX)}$  vs. Copper Area (CSP4)

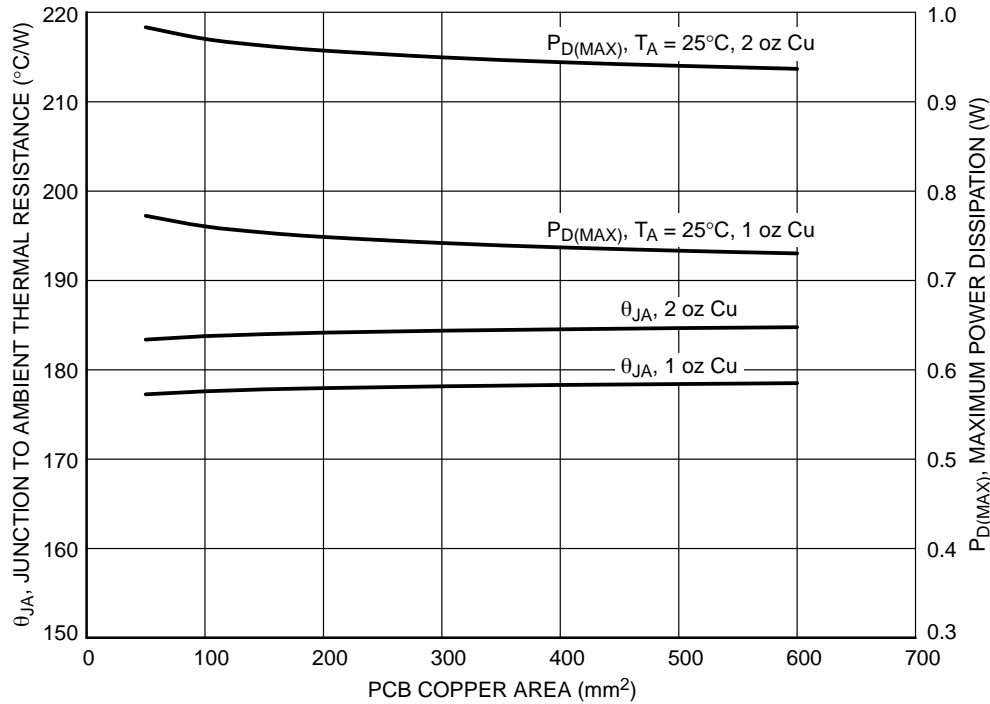


Figure 49.  $\theta_{JA}$  and  $P_{D(MAX)}$  vs. Copper Area (XDFN4)

## NCP163

### Reverse Current

The PMOS pass transistor has an inherent body diode which will be forward biased in the case that  $V_{OUT} > V_{IN}$ . Due to this fact in cases, where the extended reverse current condition can be anticipated the device may require additional external protection.

### Power Supply Rejection Ratio

The NCP163 features very high Power Supply Rejection ratio. If desired the PSRR at higher frequencies in the range 100 kHz – 10 MHz can be tuned by the selection of  $C_{OUT}$  capacitor and proper PCB layout.

### Turn-On Time

The turn-on time is defined as the time period from EN assertion to the point in which  $V_{OUT}$  will reach 98% of its nominal value. This time is dependent on various application conditions such as  $V_{OUT(NOM)}$ ,  $C_{OUT}$ ,  $T_A$ .

### PCB Layout Recommendations

To obtain good transient performance and good regulation characteristics place  $C_{IN}$  and  $C_{OUT}$  capacitors close to the device pins and make the PCB traces wide. In order to minimize the solution size, use 0402 or 0201 capacitors with appropriate capacity. Larger copper area connected to the pins will also improve the device thermal resistance. The actual power dissipation can be calculated from the equation above (Equation 2). Expose pad can be tied to the GND pin for improvement power dissipation and lower device temperature.

# NCP163

## ORDERING INFORMATION (WLCSP4)

Device	Voltage Option	Marking	Rotation	Description	Package	Shipping†
NCP163AFCS180T2G	1.8 V	Y	180	250 mA, Active Discharge	WLCSP4 CASE 567KA (Pb-Free)	5000 / Tape & Reel
NCP163AFCS260T2G	2.6 V	4	180			
NCP163AFCS280T2G	2.8 V	3	180			
NCP163AFCS285T2G	2.85 V	5	180			
NCP163AFCS290T2G	2.9 V	6	180			
NCP163AFCS2925T2G	2.925 V	2	180			
NCP163BFCS180T2G	1.8 V	Y	270	250 mA, Non-Active Discharge		
NCP163BFCS2925T2G	2.925 V	2	270			
NCP163AFCT180T2G	1.8 V	Y	180	250 mA, Active Discharge	WLCSP4 CASE 567JZ (Pb-Free)	5000 / Tape & Reel
NCP163AFCT260T2G	2.6 V	6	270			
NCP163AFCT280T2G	2.8 V	3	180			
NCP163AFCT285T2G	2.85 V	5	270			
NCP163AFCT290T2G	2.9 V	4	270			
NCP163AFCT2925T2G	2.925 V	2	180			
NCP163AFCT300T2G	3.0 V	3	270	250 mA, Non-Active Discharge		
NCP163BFCT180T2G	1.8 V	Y	270			
NCP163BFCT2925T2G	2.925 V	2	270			

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

## ORDERING INFORMATION (XDFN4)

Device	Voltage Option	Marking	Description	Package	Shipping†
NCP163AMX120TBG*	1.2 V	ME	250 mA, Active Discharge	XDFN4 CASE 711AJ (Pb-Free)	3000 / Tape & Reel
NCP163AMX130TBG*	1.3 V	MG			
NCP163AMX180TBG	1.8 V	MA			
NCP163AMX1825TBG	1.825 V	MC			
NCP163AMX190TBG	1.9 V	MH			
NCP163AMX260TBG	2.6 V	MN			
NCP163AMX275TBG	2.75 V	MD			
NCP163AMX280TBG	2.8 V	MM			
NCP163AMX285TBG	2.85 V	MQ			
NCP163AMX290TBG	2.9 V	MR			
NCP163AMX300TBG	3.0 V	MJ			
NCP163AMX330TBG	3.3 V	MK			
NCP163AMX500TBG	5.0 V	ML			
NCP163BMX180TBG	1.8 V	PA	250 mA, Non-Active Discharge		
NCP163BMX1825TBG	1.825 V	PC			
NCP163BMX275TBG	2.75 V	PD			

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

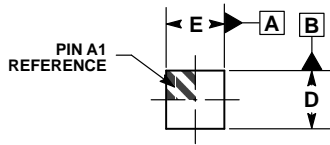
\*Contact sales office for availability information.



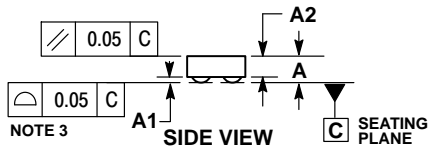
# NCP163

## PACKAGE DIMENSIONS

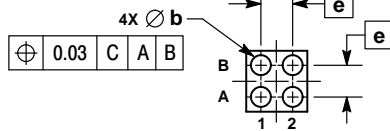
### WLCSP4, 0.64x0.64 CASE 567JZ ISSUE A



TOP VIEW



SIDE VIEW



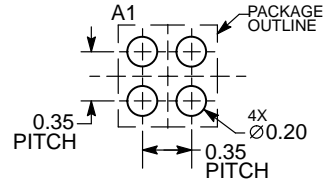
BOTTOM VIEW

NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. COPLANARITY APPLIES TO SPHERICAL CROWNS OF SOLDER BALLS.

MILLIMETERS			
DIM	MIN	NOM	MAX
A	---	---	0.33
A1	0.04	0.06	0.08
A2	0.23 REF		
b	0.195	0.210	0.225
D	0.610	0.640	0.670
E	0.610	0.640	0.670
e	0.35 BSC		

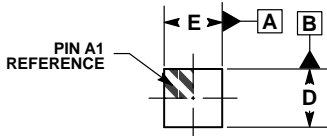
### RECOMMENDED SOLDERING FOOTPRINT\*



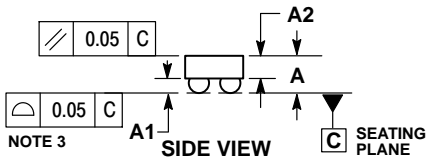
DIMENSIONS: MILLIMETERS

\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

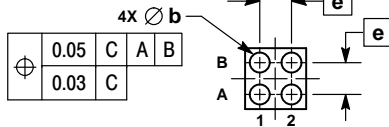
### WLCSP4, 0.64x0.64 CASE 567KA ISSUE A



TOP VIEW



SIDE VIEW



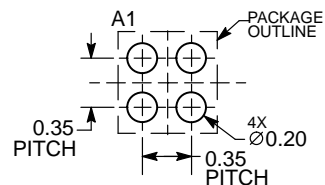
BOTTOM VIEW

NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. COPLANARITY APPLIES TO SPHERICAL CROWNS OF SOLDER BALLS.

MILLIMETERS			
DIM	MIN	NOM	MAX
A	0.35	0.40	0.45
A1	0.14	0.16	0.18
A2	0.25 REF		
b	0.185	0.200	0.215
D	0.610	0.640	0.670
E	0.610	0.640	0.670
e	0.35 BSC		

### RECOMMENDED SOLDERING FOOTPRINT\*



DIMENSIONS: MILLIMETERS

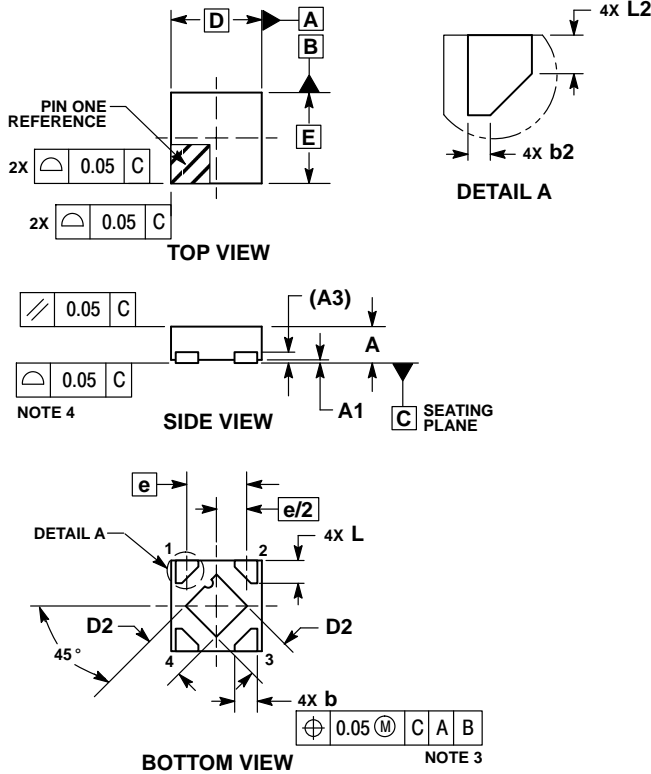
\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.



# NCP163

## PACKAGE DIMENSIONS

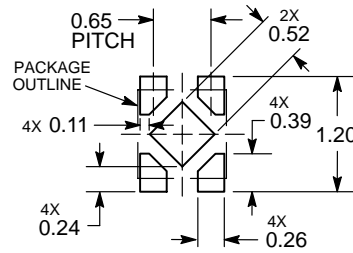
### XDFN4 1.0x1.0, 0.65P CASE 711AJ ISSUE A



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
  2. CONTROLLING DIMENSION: MILLIMETERS.
  3. DIMENSION b APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.15 AND 0.20 mm FROM THE TERMINAL TIPS.
  4. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.

MILLIMETERS		
DIM	MIN	MAX
A	0.33	0.43
A1	0.00	0.05
A3	0.10 REF	
b	0.15	0.25
b2	0.02	0.12
D	1.00 BSC	
D2	0.43	0.53
E	1.00 BSC	
e	0.65 BSC	
L	0.20	0.30
L2	0.07	0.17

### RECOMMENDED MOUNTING FOOTPRINT\*



DIMENSIONS: MILLIMETERS

\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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